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NUMERO: 2326A

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A P P U N T I

STUDENTE: Caldera Filippo

MATERIA: Estimation Filtering and System Identification - Prof. Taragna

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**ATTENZIONE: QUESTI APPUNTI SONO FATTI DA STUDENTIE NON SONO STATI VISIONATI DAL DOCENTE.
IL NOME DEL PROFESSORE, SERVE SOLO PER IDENTIFICARE IL CORSO.**

EF SI

ESTIMATION, FILTERING
AND SYSTEM IDENTIFICATION

- M. TARAGNA -

FILIPPO
CALDERA

So...

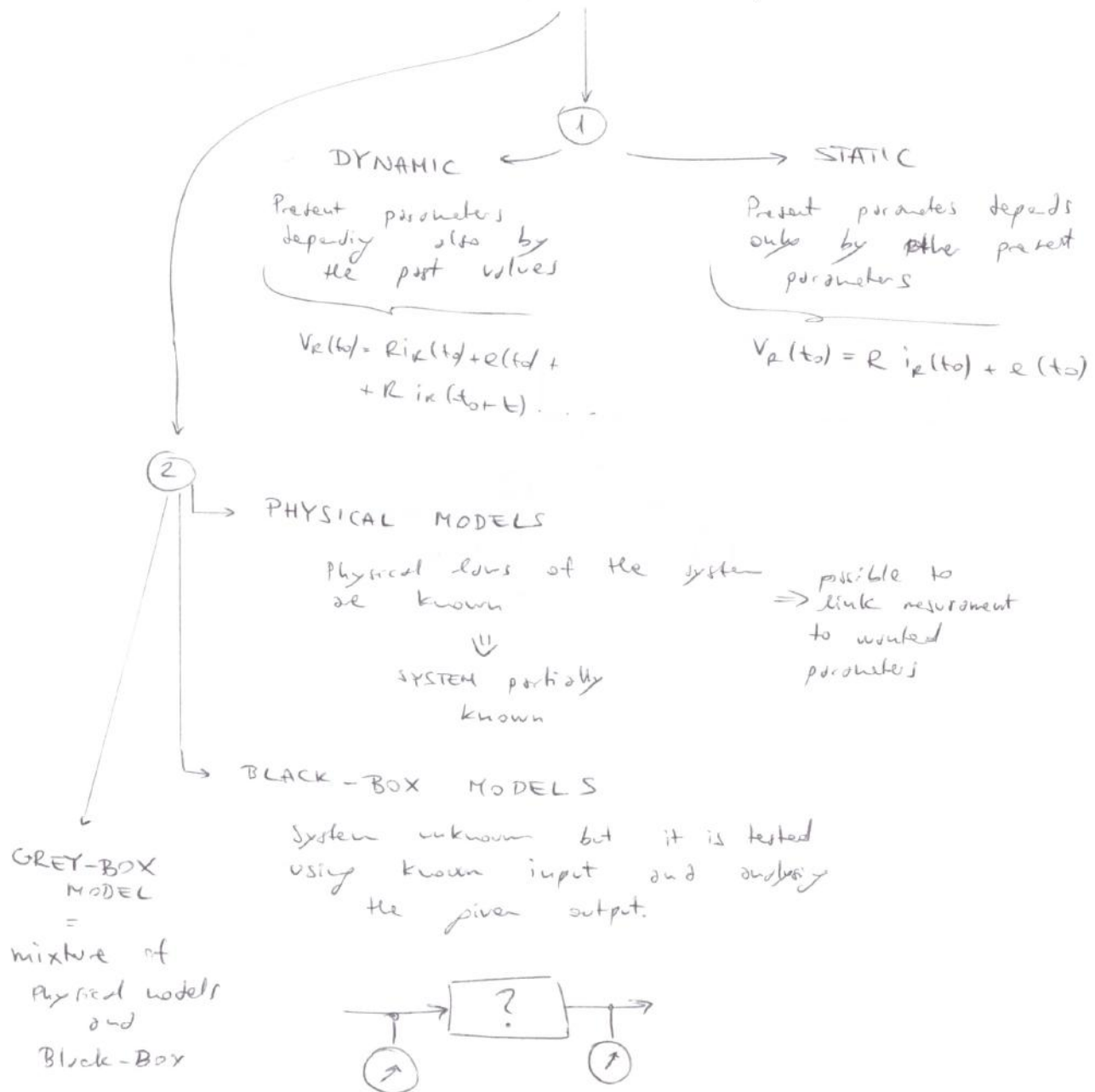
$$V_R = R i_R + e$$

Assuming model exist, characterized by some parameters and noise too.

All starting by data

Model obtained is a mathematical model, ruled by parameters.

There are two main distinction categories



Basing of θ behavior we can classify different estimation method

$\theta(t) = \theta = \text{const} \Rightarrow$ PARAMETRIC IDENTICAL PROBLEM

estimate indicated $\hat{\theta}$ or $\hat{\theta}_T$

• Need to find something else to something unknown.

$\theta(t) \neq \text{const}$
 $= \hat{\theta}(t|N)$

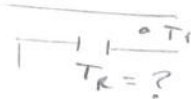
\Rightarrow estimator $\hat{\theta}(t|T), \hat{\theta}(t|N)$

$t > t_N \Rightarrow$ PREDICTION PROBLEM
 Trying to figure out what would be the value in the future

$t = t_N \Rightarrow$ FILTERING PROBLEM

Used where estimate is not measurable or is remote measure so it is estimated by measuring something else

ES. Room Temperature but sensor in the hallway



$t_1 < t < t_N \Rightarrow$ REGULATION / INTERPOLATION PROBLEM
 (SMOOTHING PROBLEM)

Trying to estimate the parameter values for lines between the measured one



Quality of the estimation is given by how $\hat{y}(t)$ is close to $y(t)$



Quality of estimation given by prediction error

$$\varepsilon(t) = y(t) - \hat{y}(t)$$

if $\varepsilon \rightarrow 0 \Rightarrow$ GOOD PREDICTION

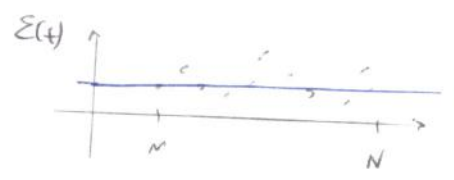
There are other additional methods like

min-MAX algorithm \rightarrow based on cost function

$$J(\theta) = \max |\varepsilon(x)|$$

$$\hat{\theta}_{\text{optimal}} = \min (J(\theta)) = \min \{ \max \{ |\varepsilon(x)| \} \}$$

Best criteria to be applied is given by the prediction error behavior

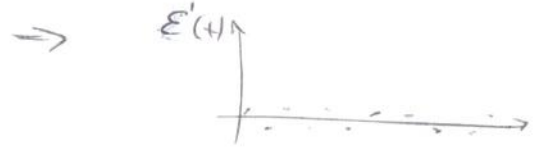


$E[\varepsilon(t)] \neq 0 \Rightarrow E[\varepsilon(t)] \neq 0$ means $\varepsilon(t)$ has BIAS

To correct the BIAS a constant term is added and considered in the estimation process

↓
Something in the system has been neglected

$$\hat{y}(t) = \textcircled{\theta_0} + \theta_1 y(t-1) + \dots$$



$E[\varepsilon'(t)] = 0$ ← Solved BIAS

HOW TO DESCRIBE A RANDOM VARIABLE:

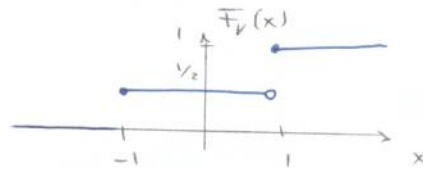
There are a large set of tools to describe a Random V.

• PROBABILITY DISTRIBUTION FUNCTION $F_V(x)$

$$F_V(x) = P(V \leq x)$$

- $F(-\infty) = 0$
- $F(+\infty) = 1$
- $F(x_1) \leq F(x_2) \quad \forall x_1 < x_2$

es.

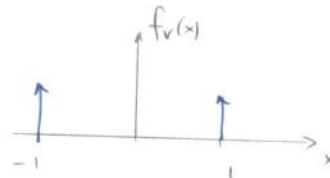
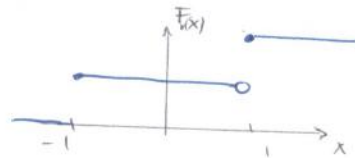


• PROBABILITY DENSITY FUNCTION $f_V(x)$

$$f_V(x) = \frac{d}{dx} F_V(x)$$

- $f(x) \geq 0 \quad \forall x$
- $\int_{-\infty}^{+\infty} f_V(x) dx = 1$
- $F_V(x) = \int_{-\infty}^x f_V(\xi) d\xi$

es.



• K^{th} - ORDER MOMENTUM

$$m_k[V] = E[V^k] = \int_{-\infty}^{+\infty} x^k f_V(x) dx$$

$$m_0[V] = E[V] = \mu$$

• K^{th} - ORDER CENTRAL MOMENTUM

$$\mu_k[V] = E[(V - E[V])^k] = \int_{-\infty}^{+\infty} (x - E[V])^k f_V(x) dx$$

$$\mu_0[V] = 0$$

$$\mu_2[V] = \sigma_V^2$$

Sometimes multitude of Random Variables are required



Better use Random Variables Vector

• MEAN / EXPECTED VALUE

$$\begin{cases} E[U] = [E[U_1], \dots, E[U_n]]^T \\ E[U_i] = \int_{-\infty}^{+\infty} x_i f_i(x) dx \end{cases}$$

• VARIANCE MATRIX Σ_V

$$\begin{aligned} \Sigma_V = V[U] &= E[(U - E[U])(U - E[U])^T] = \\ &= \int_{\mathbb{R}^n} (x - E[U])(x - E[U])^T f(x) dx \end{aligned}$$

PROPERTIES

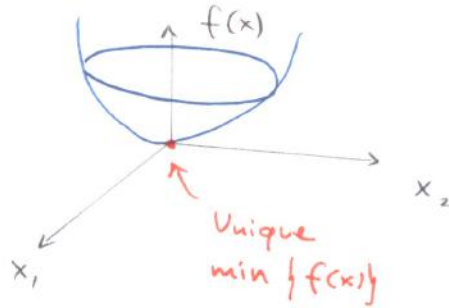
- Σ_V symmetric $\Rightarrow \Sigma_V = \Sigma_V^T$
- Positive semidefinite
- $\lambda_i(\Sigma_V) = E[X^T (U - E[U])^2] \geq 0 \quad \forall x \in \mathbb{R}^n$
- $[\Sigma_V]_{ii} = \sigma_i^2 = \text{VARIANCE}$
- $[E_V]_{ij} = \sigma_{ij} = \text{COVARIANCE}$

$$\Sigma_V = \begin{bmatrix} \sigma_1^2 & \sigma_{12} & \dots & \dots \\ \sigma_{21} & \sigma_2^2 & \dots & \dots \\ \dots & \dots & \dots & \dots \\ \dots & \dots & \dots & \sigma_n^2 \end{bmatrix}$$

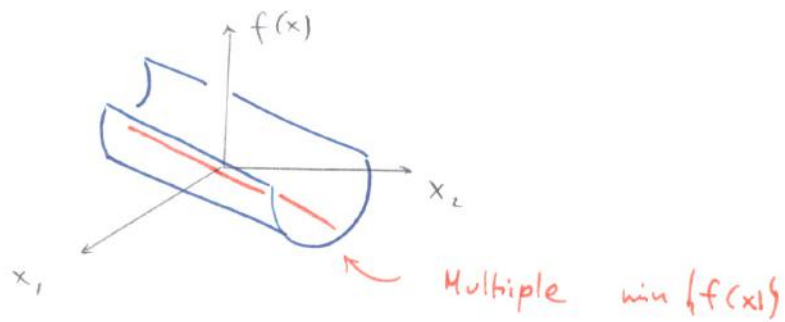
Anyway if $\det(\Sigma_v) \geq 0$ that means

• $\det(\Sigma_v) > 0 \iff \lambda_i > 0 \forall i$

Case $N=2$



• $\det(\Sigma_v) = 0 \iff \exists i \mid \lambda_i = 0$



if $\rho_{ij} = 0 \Rightarrow i, j$ UNCORRELATED $\not\Rightarrow$ INDEPENDENT

Having covariance matrix it's possible to "normalize it" to get NORMALIZED COVARIANCE MATRIX P_V

$$P_V = \begin{bmatrix} \rho_{11} & \rho_{12} & \dots \\ & \rho_{22} & \dots \\ & & \dots \\ & & & \rho_{nn} \end{bmatrix} = \begin{bmatrix} 1 & \rho_{12} & \dots \\ & 1 & \dots \\ & & \dots \\ & & & 1 \end{bmatrix}$$

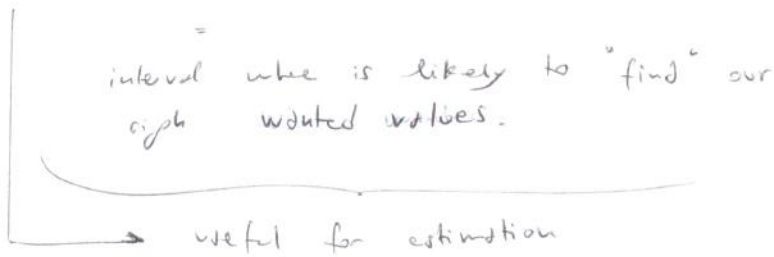
P_V has the same properties of Σ_V

we have a relevant case:

$$\Sigma_w = \begin{bmatrix} \sigma_1^2 & & 0 \\ & \ddots & \\ 0 & & \sigma_n^2 \end{bmatrix} \Rightarrow P_w = \begin{bmatrix} 1 & & 0 \\ & 1 & \\ 0 & & \ddots \end{bmatrix}$$

In our application $P[a < x < b]$ is

CONFIDENCE INTERVAL



Since our measurement are affected by noise (which is considered a R.V.) then measurement itself is a R.V.



ESTIMATE IS A R.V.
NOT A VALUE



To have an idea where estimate real value is located we use confidence interval.

NORMAL/GAUSSIAN USING VECTORS

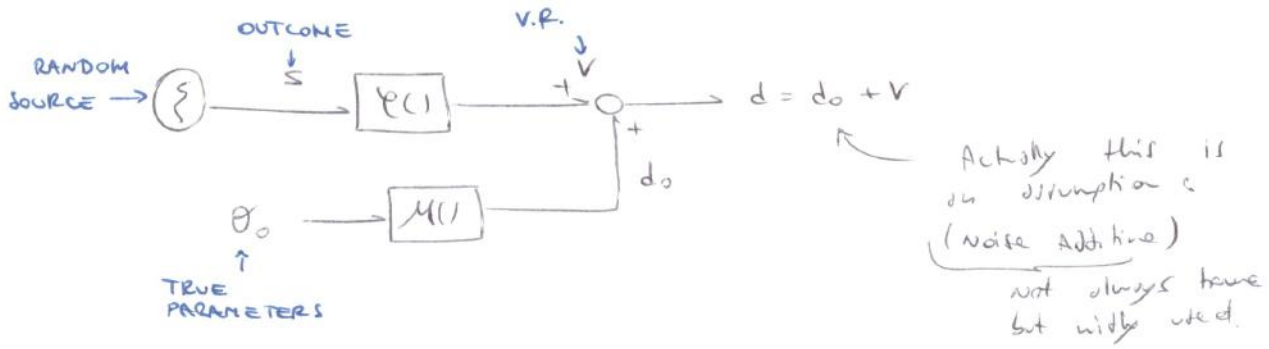
if $f(x) = \frac{1}{(2\pi)^{n/2} \sqrt{\det \Sigma}} e^{-\frac{1}{2}(x-\bar{x})^T \Sigma^{-1}(x-\bar{x})} \Rightarrow$ vector is Jointly Gaussian

PROPERTIES

If $v_1, \dots, v_n \sim N \Rightarrow$ they are also Jointly Normal

If $\bullet v_1, \dots, v_n$ Jointly Gaussian \Rightarrow w_1, \dots, w_n INDEPENDENT.
 $\bullet v_1, \dots, v_n$ uncorrelated

PROBABILISTIC DATA DESCRIPTION



$$\hat{\theta}_0 = f(d) = \text{estimation result}$$

(since $d = d_0 + V$, $V = \text{R.V.} \Rightarrow d$ is R.V.)

In real application it's not true that $\hat{\theta}_0 \approx \theta_0$

We wish that, but how to be sure $\hat{\theta}_0 \approx \theta_0$ if θ_0 is unknown?

Some tricks needed.

=
 Define some estimation characteristics to be almost true of θ_0 and

An interesting point is ALMOST-SURE-CONVERGENCE

when $N = \text{high}$ there are more chances to have bad data point.

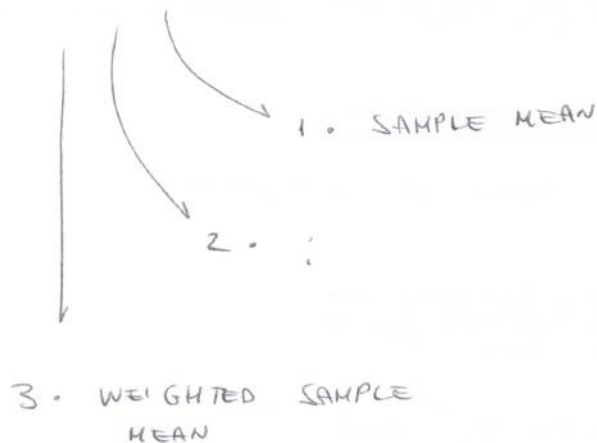
If # bad data point is negligible \Rightarrow Possible to neglect them

But it's fine anyway
It's called ALMOST-SURE CONVERGENCE

Not considering all data point
and convergence is more sure

After all, data point deleted are so few.

Some estimator examples



important

Important fact is σ^2 can't be less than a certain value.

So, when analyzing data affected by noise is NOT possible to get noise = 0 but

$$\sigma_{\hat{\theta}}^2 \geq m^{-1}$$

CRAMER-RAO
INEQUALITY



m = Fisher information quantity

depending on assumption made.

Best case possible is $\sigma_{\hat{\theta}}^2 = m^{-1}$

In case of $\theta \in \mathbb{R}^m$ we have

$$\text{Var}[\hat{\theta}] \geq M^{-1}$$

that means

Fisher Information Matrix

$$\text{Var}[\hat{\theta}] - M^{-1} \geq 0 \iff \forall x \in \mathbb{R}^m \quad x^T (\text{Var}[\hat{\theta}] - M^{-1}) x \geq 0$$



$$\{\text{Var}[\hat{\theta}] - M^{-1}\}_{ii} \geq 0$$

\iff $\text{Var}[\hat{\theta}]$ is matrix positive semi-definite

$$\{\text{Var}[\hat{\theta}]\}_{ii} - \{M^{-1}\}_{ii} \geq 0$$

$$\{\text{Var}[\hat{\theta}]\}_{ii} \geq \{M^{-1}\}_{ii}$$

σ^2 ↑ Fisher quantity

as scalar, σ^2 is low bounded by Fisher quantities

this form $Y = \phi \cdot \theta$ is not so true due noise

correct one is: $Y \approx \phi \theta$

$Y \approx \phi \theta$ ← unknown ⇒ to get θ then ϕ must be invertible

- ⇓
- $\phi \in \mathbb{R}^{N \times m}$ ($N = m$)
- $\det(\phi) \neq 0$

not always possible

In most cases ϕ is not squared
($N \neq m$)

we need to solve:

- find way not to get ϕ^{-1} or been able to obtain θ
- evaluate how good estimation is

} All problems solved in 'one shot'.

Ability to measure the distance between estimate value and the measured one.

$$\| \underbrace{\phi \theta}_{\substack{\text{output} \\ \text{computed using} \\ \theta}} - \underbrace{Y}_{\substack{\text{measured} \\ \text{output}}} \|_2^2$$

$$\| \phi \theta - Y \|_2^2 = \sum_{k=1}^N \left[\underbrace{(\phi \theta)_k}_{\substack{\uparrow \\ k^{\text{th}} \text{ row}}} - \underbrace{Y_k}_{\substack{\uparrow \\ k^{\text{th}} \text{ measurement}}} \right]^2 = \sum_{k=1}^N [\varphi(k)^T \theta - y(k)]^2$$

need to have it as small as possible

But... if

$$\theta = (\phi^T \phi)^{-1} \phi^T Y$$

How to compute the pseudo inverse matrix of ϕ ?

3 ways

① $\theta_{LS} = \text{pinv}(\phi) * Y$

② $\theta_{LS} = \text{inv}(\phi^T * \phi) * \phi^T * Y$ DO NOT USE

if $\det(\phi) \rightarrow 0 \neq 0$ then θ_{LS} would be a random vector \Rightarrow NO GOOD

③ $\theta_{LS} = \phi \backslash Y$

NOTE:

if ϕ squared = $N=m \Rightarrow (\phi^T \phi)^{-1} \phi^T$

if $\det(\phi^T \phi) \neq 0$
 \Downarrow
 $\det(\phi) \neq 0$
 \Downarrow
 ϕ invertible itself

$$\left(\begin{matrix} m & m \\ m & m \end{matrix} \right) \begin{matrix} m \\ m \end{matrix}$$

$$\begin{matrix} m & m \\ m & m \end{matrix}$$

$$\Rightarrow (\phi^T \phi)^{-1} = \phi^{-1} (\phi^T)^{-1}$$

$$\text{pinv}(\phi) = (\phi^{-1}) \underbrace{(\phi^T)^{-1}}_I \phi^T = \phi^{-1}$$

If $N=m \Rightarrow \phi$ squared $\Rightarrow \text{pinv}(\phi) = \text{inv}(\phi)$

PROBABILISTIC CHARACTERISTIC OF LS ESTIMATOR

Given $\hat{\theta}_{LS} = (\phi^T \phi)^{-1} \phi^T Y$

$\exists \hat{\theta}_{LS} \Leftrightarrow [\phi^T \phi]^{-1} \phi^T Y$ is invertible } MAIN ASSUMPTION

If $Y(t) = \psi(t)^T \theta_0 + v(t)$ assuming θ_0 exists

$$\begin{cases} Y = \phi \theta_0 + v \\ v = \text{R.V.} \mid E[v] = 0 \end{cases}$$



$$\begin{aligned} \hat{\theta} &= (\phi^T \phi)^{-1} \phi^T Y = (\phi^T \phi)^{-1} \phi^T (\phi \theta_0 + v) = \\ &= (\phi^T \phi)^{-1} \phi^T \phi \theta_0 + (\phi^T \phi)^{-1} \phi^T v = \\ &= \theta_0 + (\phi^T \phi)^{-1} \phi^T v \end{aligned}$$



$$E[\hat{\theta}] = E[\theta_0] + \underbrace{(\phi^T \phi)^{-1} \phi^T}_{=0} E[v] = \theta_0 \Rightarrow \text{UNBIASED}$$

If $Y = \phi \theta_0 + v$
 $v = \text{R.V.} \mid E[v] = 0 \Rightarrow$ LS estimator UNBIASED

NOTE:

In general $\hat{\theta}_{LS} \neq \theta_0$ but

if $v \in \text{Kern}(\phi) \Rightarrow \hat{\theta}_{LS} = \theta_0$ } very rare

Then there is also another issue

$$\Sigma_v = \sigma_v^2 (\Phi^T \Phi)^{-1}$$

↑
unknown in
real application

to get σ_v^2 EXPERIMENTAL NOISE VARIANCE is needed

$$\sigma^2 = \frac{J(\hat{\theta}_{LS})}{N-m} = \frac{J(\hat{\theta}_{LS})}{N} = \frac{\|Y - \Phi \hat{\theta}\|_2^2}{N}$$

↑
 $N \gg m$

To obtain a better estimation in WLS there is an additional degree of freedom given by Q .

Proven $Q = \Sigma_v^{-1} \Rightarrow V[\hat{\theta}_{WLS}] = \text{min. possible}$

$$Q = \Sigma_v^{-1} = \begin{bmatrix} \sigma_1^2 & & \\ & \dots & \\ & & \sigma_N^2 \end{bmatrix}^{-1} = \begin{bmatrix} 1/\sigma_1^2 & & \\ & \dots & \\ & & 1/\sigma_N^2 \end{bmatrix}$$

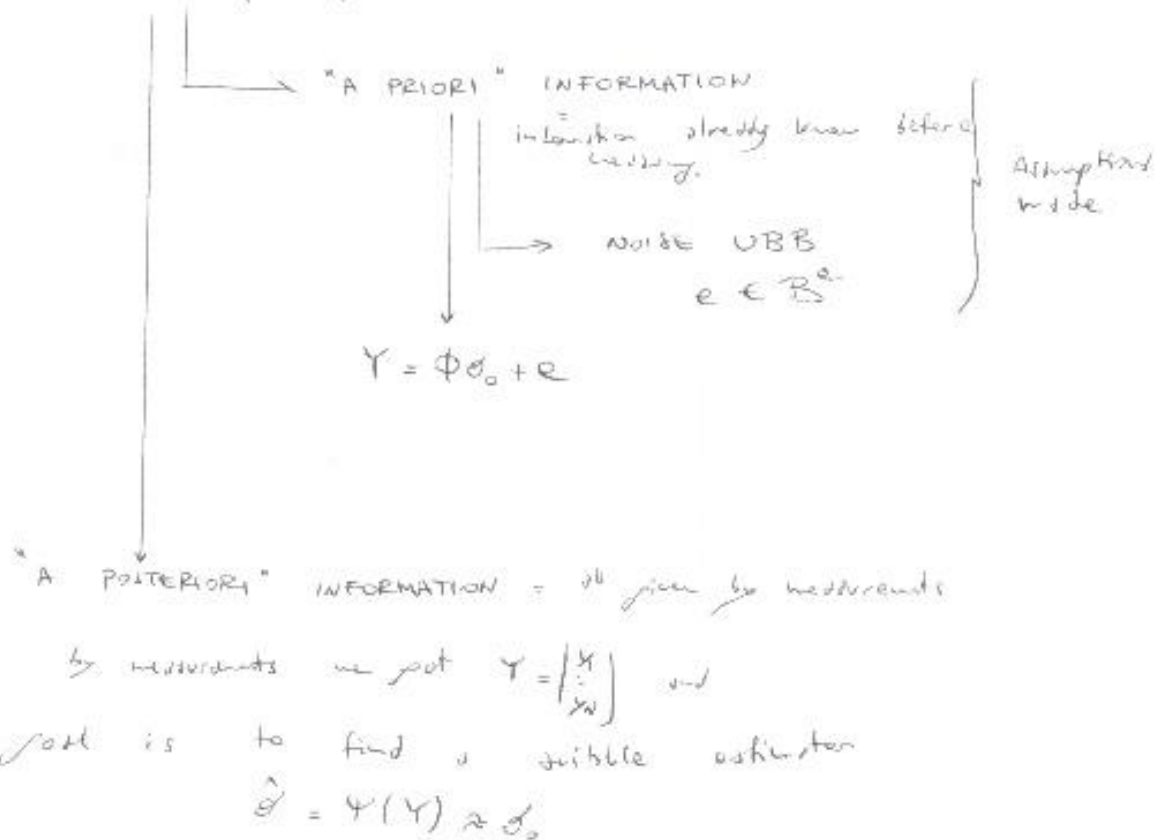
GAUSS-MARKOV ESTIMATE

$$\hat{\theta}_{GM} = [\Phi^T \Sigma_v^{-1} \Phi]^{-1} \Phi^T \Sigma_v^{-1} Y$$

$$V[\hat{\theta}_{GM}] = [\Phi^T \Sigma_v^{-1} \Phi]^{-1}$$

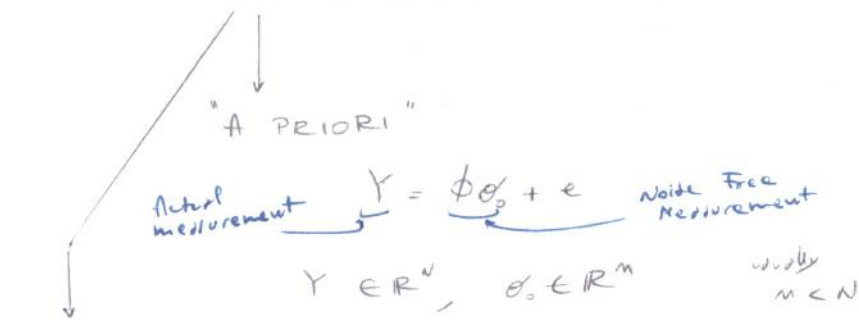
Linear estimation in set membership framework (Unknown But Bounded Framework)

Problem to be solved based in
2 kind of information



03/23/2017

LINEAR ESTIMATION PROBLEM IN THE SET MEMBERSHIP FRAMEWORK

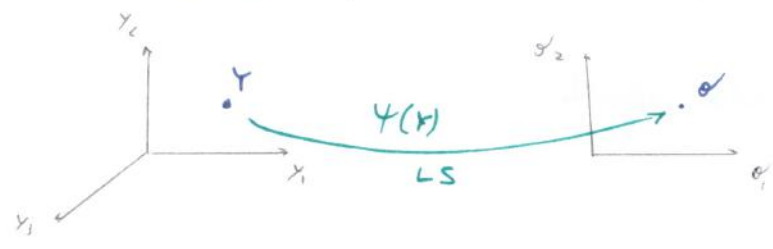


"A POSTERIORI"

All operating $\rightarrow Y$, known the measurements

EX.

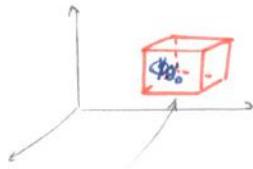
- $N = 3 \Rightarrow$ Measurement Space = \mathbb{R}^3
- $m = 2 \Rightarrow$ Parameter Space = \mathbb{R}^2



Where

$$\psi(Y) = \underset{\text{LS}}{\underbrace{(\phi^T \phi)^{-1}}_{\uparrow}} \phi^T Y = \text{pinv}(\phi) Y = A_{LS} Y$$

In space is just a mapping transformation between different spaces.



cube is MUS
Measurement Uncertainty Set

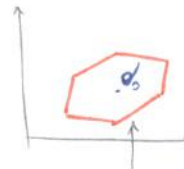
$$MUS = y \oplus \mathcal{B}_e^0 = \{ \tilde{y} \mid y - \tilde{y} \in \mathcal{B}_e^0 \}$$

↑
Error

FOR SURE

$$\phi \theta_0 \in MUS^0$$

Real output value is inside the cube



MUS mapping generates

EUS
Estimate Uncertainty Set

To have $\theta_0 \in EUS^T$ on entire MUS mapping is required

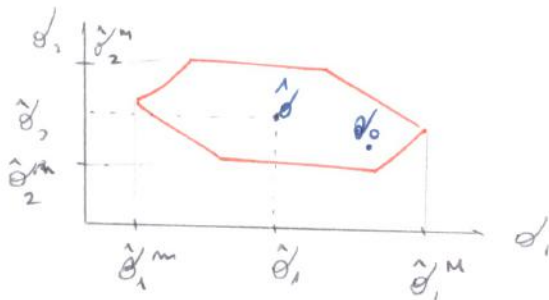
If $\phi \theta_0$ is mapped then θ_0 is found.

$$A_{LS} \phi \theta_0 = (\phi^T \phi)^{-1} \phi^T \phi \theta_0 = \theta_0$$

Real θ_0 is inside EUS

$\theta_0 \in EUS$ FOR SURE (TOO)

Since, starting from Y , \tilde{y} is founded we can use ESTIMATE UNCERTAINTY INTERVALS (EUI) to better understand how θ_0 components are near to EUS border



EUI just a projection of EUS!!

Noise is URB \Rightarrow Rounded by components

IDEA!

Focus on cube's vertices and mapping them.

In EUS just connect the dots

but

vertices of MUS = $2^N \Rightarrow$ If $N \uparrow \Rightarrow$ MAPPING WOULD BE A MESS

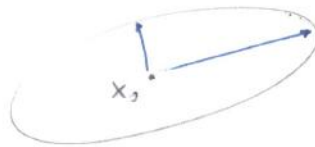
\Downarrow

NOT USED IN PRACTICE

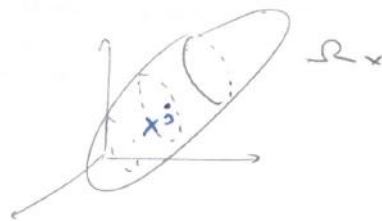
If Ω_x is ellipsoid centered in x_0



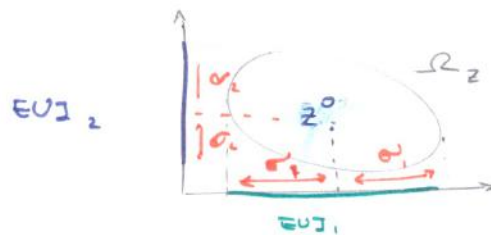
- main axes direction given by eigenvectors of Σ_x
- semi-axes lengths given by $\sqrt{\lambda_i(\Sigma)}$



If



and $\text{rank}(A_{LS}) = m$



$$z^0 = A_{LS} x^0$$

σ_j evaluated using $\sigma_j = \sqrt{[(\Phi^T \Phi)^{-1}]_{jj}}$

To estimate how far y to $\phi\tilde{\theta}$ is we need to compute the distance using linear \downarrow .

$$|(y - \phi\tilde{\theta})_i| = |y_i - \varphi_i^T \tilde{\theta}| \leq \varepsilon$$

$$\Downarrow$$

$$-\varepsilon \leq y_i - \varphi_i^T \tilde{\theta} \leq \varepsilon$$

$$\underbrace{\hspace{10em}}_{\text{putting in vertices}}$$

$$\varphi_i^T \tilde{\theta} \leq y_i + \varepsilon \qquad \qquad \qquad -\varphi_i^T \tilde{\theta} \leq \varepsilon - y_i$$

$$\begin{bmatrix} \varphi_1^T \\ \vdots \\ \varphi_N^T \\ -\varphi_1^T \\ \vdots \\ -\varphi_N^T \end{bmatrix} \tilde{\theta} \leq \begin{bmatrix} y_1 \\ \vdots \\ y_N \\ -y_1 \\ \vdots \\ -y_N \end{bmatrix} + \varepsilon$$

$$H \tilde{\theta} \leq b \quad \leftarrow \text{matrix linear inequalities}$$

Solution by linear programming

$$\begin{cases} \min \{c^T \tilde{\theta}\} \\ c^T = \text{lower} \in \mathbb{R}^{1 \times m} \end{cases}$$

using MATLAB

$$\text{linprog}(c^T, \dots)$$

$$c^T = [0, \dots, \underset{\uparrow j^{\text{th}}}{1}, \dots, \underset{\uparrow N^{\text{th}}}{\sigma}]$$

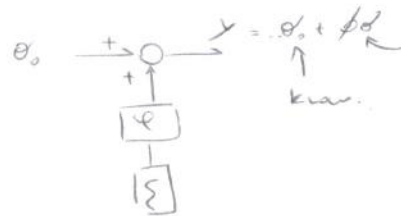
Return Parameter Uncertainty Intervals [PUI]

$$\text{PUI} = [\min \theta_j, \max(\theta_j)] = [\theta_j^m, \theta_j^M]$$

MAXIMUM LIKELIHOOD ESTIMATORS

03/30/2017

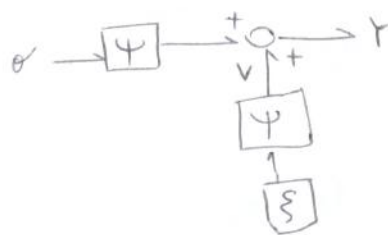
In the classical approach we had supposed θ_0 as
 a known-function
 So, the idea was



NEW APPROACH is considering θ_0 as a deterministic value instead of θ

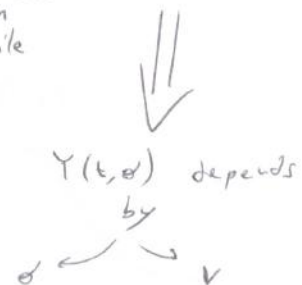


No more fixed θ_0



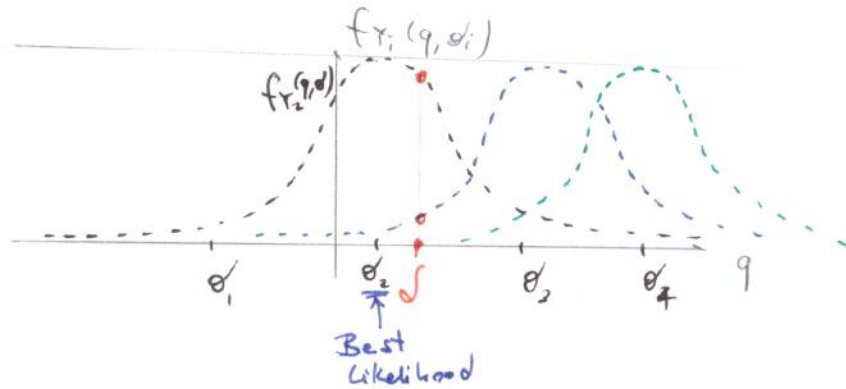
$$y(t) = \psi(t, \theta) + v(t) = \varphi(t, \theta)$$

↑
Random Variable



$$L(\theta) = f_Y(q, \theta) \Big|_{q=d} \quad d \text{ is measurement}$$

Since d is fixed due measure, then with multiple θ_i we get



Best value depends by θ_i , best one is θ_2 because is the closest one to d and is more likely to be near the measure.

$f_{Y_2}(q, \theta_2)$ is MAXIMUM LIKELIHOOD ESTIMATION

$$\theta_2 = \hat{\theta}_{ML} = \arg \max_{\theta \in \mathbb{R}} \{L(\theta)\}$$

PROPERTIES

$$E[\hat{\theta}_{ML}] \xrightarrow{N \rightarrow \infty} \theta$$

$$\sum_{\theta_{ML}} \leq \sum_{\theta} \quad N \rightarrow \infty \quad \left. \begin{array}{l} \text{EFFICIENCY} \\ \text{Reduces uncertainties} \end{array} \right\}$$

$$\theta_{ML} \sim N(\theta, \sigma^2) \quad N \rightarrow \infty \quad \left. \begin{array}{l} \text{Asymptotically Gaussian} \\ \Downarrow \end{array} \right\}$$

If noisy measurements for $N \rightarrow \infty$ we just a Gaussian signal

EX. LINEAR CASE

$$Y(t, \theta) = \Psi'(t) \theta \Rightarrow Y(\theta) = \begin{bmatrix} \Psi'(t_1) \theta \\ \vdots \\ \Psi'(t_N) \theta \end{bmatrix} = \begin{bmatrix} \Psi'(t_1) \\ \vdots \\ \Psi'(t_N) \end{bmatrix} \theta = \Phi \theta$$

$$\hat{\theta}_{ML} = \underset{\theta}{\text{arg}} \left\{ \min_{\theta} \left[\underbrace{(f - \Phi \theta)^T}_{\text{Estimation error}} \Sigma_V^{-1} \underbrace{(f - \Phi \theta)}_{\text{Estimation error}} \right] \right\} \Rightarrow \begin{matrix} \text{(Quadratic function)} \\ \text{Exists unique} \\ \text{min} \end{matrix}$$

$$\hat{\theta}_{ML} = \hat{\theta}_{GM} = \text{GAUSS-MARKOV} = \text{WEIGHTED LS.}$$

NOTE!

If ϕ linear $\Rightarrow \phi$ convex $\Rightarrow \exists!$ min(ϕ)

If $\Sigma_V = \sigma_v^2 I_N$
independent, identically distributed noise $\Rightarrow \hat{\theta}_{ML} = \hat{\theta}_{LS}$

If Ψ LINEAR \Rightarrow

GAUSS-MARKOV is

- $E[\hat{\theta}_{GM}] = \theta$
- $\Sigma_{\hat{\theta}_{GM}} \leq \Sigma_{\hat{\theta}} \quad \forall \hat{\theta}$
- $\lim_{N \rightarrow \infty} \Sigma_{\hat{\theta}_{GM}} = 0$
- Gauss or

GAUSS MARKOV IS BEST ONE
IF ONLY IF ASSUMPTIONS ARE
SATISFIED



Bayesian Estimate = $\hat{\theta} = E[\theta | d = \hat{d}]$

Generic Measurement \nearrow \nwarrow actual measurement

EX. GAUSSIAN CASE (EASIEST CASE)

$\theta, d \sim N()$ and Jointly Gaussian

$\theta \in \mathbb{R} \quad d \in \mathbb{R}$

$\mu_\theta = 0 \quad \mu_d = 0$

$\searrow \quad \swarrow$ then

$$\begin{bmatrix} d \\ \theta \end{bmatrix} \sim N \left(\begin{bmatrix} 0 \\ 0 \end{bmatrix}, \Sigma = V \begin{bmatrix} d \\ \theta \end{bmatrix} = \begin{bmatrix} \sigma_{dd} & \sigma_{d\theta} \\ \sigma_{\theta d} & \sigma_{\theta\theta} \end{bmatrix} \right)$$

$\sigma_{dd} = \sigma_d^2 = V[d]$
 $\sigma_{\theta\theta} = \sigma_\theta^2 = V[\theta]$

$$f(d, \theta) = c e^{-\frac{1}{2} [d \ \theta] \Sigma^{-1} [d \ \theta]^T}$$

$$\Sigma^{-1} = \begin{bmatrix} \sigma_{dd} & \sigma_{d\theta} \\ \sigma_{\theta d} & \sigma_{\theta\theta} \end{bmatrix}$$

$$\Rightarrow \Sigma^{-1} = \frac{1}{\det \Sigma} \begin{bmatrix} \sigma_{\theta\theta} & -\sigma_{d\theta} \\ -\sigma_{\theta d} & \sigma_{dd} \end{bmatrix}$$

$$f(d, \theta) = c e^{-\frac{1}{2\sigma_{dd}} [d, \theta] \begin{bmatrix} \sigma_{\theta\theta} & -\sigma_{d\theta} \\ -\sigma_{\theta d} & \sigma_{dd} \end{bmatrix} \begin{bmatrix} d \\ \theta \end{bmatrix}}$$

$$= c' e^{-\frac{d^2}{2\sigma_{dd}}}$$



$$MSE = V[\hat{\theta} - \theta] = E[(\hat{\theta} - \theta)^2] =$$

$$= E\left[\left(\theta - \frac{\sigma_{\theta d}}{\sigma_{dd}} d\right)^2\right] = E\left[\theta^2 - 2 \frac{\sigma_{\theta d}}{\sigma_{dd}} \theta d + \frac{\sigma_{\theta d}^2}{\sigma_{dd}^2} d^2\right] =$$

$$= E[\theta^2] - 2 E\left[\frac{\sigma_{\theta d}}{\sigma_{dd}} \theta d\right] + \frac{\sigma_{\theta d}^2}{\sigma_{dd}^2} E[d^2] =$$

$$= \sigma_{\theta\theta} - 2 \frac{\sigma_{\theta d}^2}{\sigma_{dd}} \cdot \rho_{\theta d} + \frac{\sigma_{\theta d}^2}{\sigma_{dd}^2} \sigma_{dd} = \sigma_{\theta\theta} - \frac{\sigma_{\theta d}^2}{\sigma_{dd}} = \sigma_{\theta\theta}^2 > 0$$

$$= \sigma_{\theta\theta} (1 - \rho_{\theta d}^2) \geq 0$$

Mean Square Error = MSE ≥ 0

③ d, ϑ scalar

Any p.d.f.

NON-ZERO
MEAN
VALUE

$$\mu_d = \bar{d}, \mu_\vartheta = \bar{\vartheta}$$

$$\Rightarrow \hat{\vartheta} = \bar{\vartheta} + \frac{\sigma_{\vartheta d}}{\sigma_{dd}} (d - \bar{d})$$

$$v\left(\begin{pmatrix} d \\ \vartheta \end{pmatrix}\right) = \begin{pmatrix} \sigma_{dd} & \sigma_{\vartheta d} \\ \sigma_{d\vartheta} & \sigma_{\vartheta\vartheta} \end{pmatrix}$$

Then it $d' = d - \bar{d}, \vartheta' = \vartheta - \bar{\vartheta}$

$$\begin{aligned} \sigma_{d'd'} &= E[(d' - E[d'])^2] = E[(d')^2] = E[(d - \bar{d})^2] = \\ &= E[d^2 - 2d\bar{d} + \bar{d}^2] = E[d^2] - 2E[d\bar{d}] + \bar{d}^2 = \\ &= E[d^2] - 2\bar{d}E[d] + \bar{d}^2 = E[d^2] - 2\bar{d}\bar{d} + \bar{d}^2 = \\ &= E[d^2] - \bar{d}^2 = E[d^2] - E[d]^2 = \sigma_{dd} \end{aligned}$$

$$\Rightarrow \sigma_{d'd'} = \sigma_{dd}$$

$$\begin{aligned} \sigma_{\vartheta'\vartheta'} &= E[(\vartheta' - E[\vartheta'])^2] = E[(\vartheta')^2] = E[(\vartheta - \bar{\vartheta})^2] = \\ &= E[\vartheta^2 - 2\vartheta\bar{\vartheta} + \bar{\vartheta}^2] = E[\vartheta^2] - 2E[\vartheta\bar{\vartheta}] + \bar{\vartheta}^2 = \\ &= E[\vartheta^2] - 2\bar{\vartheta}E[\vartheta] + \bar{\vartheta}^2 = E[\vartheta^2] - 2\bar{\vartheta}\bar{\vartheta} + \bar{\vartheta}^2 = \\ &= E[\vartheta^2] - \bar{\vartheta}^2 = E[\vartheta^2] - E[\vartheta]^2 = \sigma_{\vartheta\vartheta} \end{aligned}$$

$$\Rightarrow \sigma_{\vartheta'\vartheta'} = \sigma_{\vartheta\vartheta}$$

$$\left\{ \begin{aligned} \hat{\vartheta} &= \frac{\sigma_{\vartheta d'}}{\sigma_{d'd'}} d' \Rightarrow \hat{\vartheta} = \frac{\sigma_{\vartheta d}}{\sigma_{dd}} (d - \bar{d}) = (\hat{\vartheta} - \bar{\vartheta}) \\ \sigma_{d'\vartheta'} &= \sigma_{d\vartheta} \\ \sigma_{\vartheta'd'} &= \sigma_{\vartheta d} \end{aligned} \right.$$

$$\hat{\vartheta} = \bar{\vartheta} + \frac{\sigma_{\vartheta d}}{\sigma_{dd}} (d - \bar{d})$$

$$\textcircled{2} \quad \hat{\theta} = \hat{\theta}_{\text{POST}} = \hat{\theta}_{\text{PRIOR}} + \frac{\sigma_{\theta d}}{\sigma_{dd}} (d - \bar{d})$$

If d affected by high uncertainties $= \sigma_{dd}$ high $\Rightarrow \hat{\theta}$ depends mostly by $\hat{\theta}_{\text{PRIOR}}$

If σ_{dd} small $\Rightarrow \hat{\theta}$ is corrected by $\frac{\sigma_{\theta d}}{\sigma_{dd}} (d - \bar{d})$ terms

$\textcircled{3}$

Posteriori uncertainty = Estimator error variance

$$\begin{aligned} V[\theta - \hat{\theta}] &= E[(\theta - \hat{\theta})^2] = \sigma_{\theta\theta} - \frac{\sigma_{\theta d}^2}{\sigma_{dd}} = \sigma_{\theta\theta} \left(1 - \frac{\sigma_{\theta d}^2}{\sigma_{\theta\theta} \sigma_{dd}}\right) = \\ &= \sigma_{\theta\theta} (1 - \rho^2) \end{aligned}$$

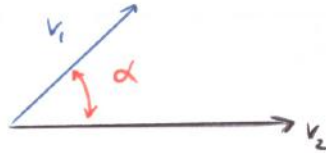
if $\rho = 0 \Rightarrow \theta, d$ uncorrelated $\Rightarrow \hat{\theta} = \hat{\theta}_{\text{PRIOR}}$

if $\rho \neq 0 \Rightarrow$ posteriori uncertainty $<$ priori uncertainty

$$\langle v_1, v_2 \rangle = \|v_1\| \cdot \|v_2\| \cos \alpha$$

$$\cos \alpha = \frac{\langle v_1, v_2 \rangle}{\|v_1\| \|v_2\|} = \frac{E[v_1, v_2]}{\sqrt{V[v_1] \cdot V[v_2]}} = \rho$$

$$\rho = \cos \alpha$$



$$\rho = 0 \Rightarrow \alpha = \pm 90^\circ \Rightarrow v_1 \perp v_2$$

=

$$v_1, v_2 \text{ uncorrelated}$$

$$\rho = \pm 1 \Rightarrow \alpha = 0 \Rightarrow v_1 \parallel v_2$$

Considering scalar Gaussian code Bayes estimate is

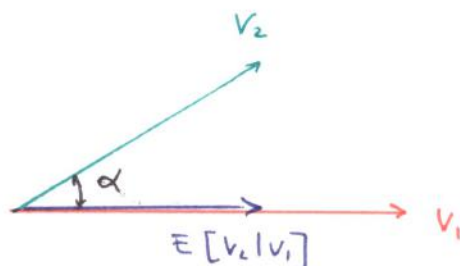
$$\hat{v}_2 = E[v_2 | v_1] = \frac{\sigma_{21}}{\sigma_{11}} v_1 = \frac{E[v_1, v_2]}{V[v_1]} v_1$$

$$\hat{v}_2 = \frac{E[v_1, v_2]}{V[v_1]} v_1 = \frac{\langle v_1, v_2 \rangle}{\|v_1\|^2} v_1 = \frac{1}{\|v_1\|} \frac{\langle v_1, v_2 \rangle}{\|v_1\| \|v_2\|} \|v_2\| v_1 =$$

$$= \|v_2\| \frac{v_1}{\|v_1\|} \cdot \cos \alpha$$

← version

⇒ \hat{v}_2 has same direction of v_1



04/10/2017

Given $\theta, d(1), d(2)$ scalar R.V. where

$$\mu_\theta = \mu_{d_1} = \mu_{d_2} = 0$$



$$\begin{bmatrix} \theta \\ d(1) \\ d(2) \end{bmatrix} \sim \left(\begin{bmatrix} 0 \\ 0 \\ 0 \end{bmatrix}, \mathbb{V} \begin{bmatrix} \theta \\ d(1) \\ d(2) \end{bmatrix} \right) \quad \text{where}$$

$$\mathbb{V} \begin{bmatrix} \theta \\ d(1) \\ d(2) \end{bmatrix} = \Sigma = \begin{bmatrix} \sigma_{\theta\theta} & \sigma_{\theta 1} & \sigma_{\theta 2} \\ \sigma_{1\theta} & \sigma_{11} & \sigma_{12} \\ \sigma_{2\theta} & \sigma_{21} & \sigma_{22} \end{bmatrix} = \begin{bmatrix} \Sigma_{\theta\theta} & \Sigma_{\theta d} \\ \Sigma_{d\theta} & \Sigma_{dd} \end{bmatrix}$$

↑ Posteriori Info

$$\Sigma_{\theta d} \text{ symmetrical} \Rightarrow \Sigma_{\theta d} = \Sigma_{d\theta}^T$$

$$\Sigma_{dd} \text{ symmetrical} \Rightarrow \Sigma_{dd} = \Sigma_{dd}^T$$

$$\Sigma_{d\theta} = \Sigma_{\theta d}^T$$



$$\begin{bmatrix} \theta \\ d(1) \\ d(2) \end{bmatrix} \sim \left(\begin{bmatrix} 0 \\ 0 \\ 0 \end{bmatrix}, \begin{bmatrix} \Sigma_{\theta\theta} & \Sigma_{\theta d} \\ \Sigma_{d\theta} & \Sigma_{dd} \end{bmatrix} \right)$$

So...

$$\begin{cases}
 E[\theta | d(1), d(2)] = \Sigma_{\theta d} \Sigma_{dd}^{-1} d \\
 \Sigma_{\theta d} = [\sigma_{\theta_1} \quad \sigma_{\theta_2}] \\
 \Sigma_{dd}^{-1} = \frac{1}{\sigma_{11}\sigma_{22} - \sigma_{12}^2} \begin{bmatrix} \sigma_{22} & -\sigma_{12} \\ -\sigma_{12} & \sigma_{11} \end{bmatrix} \\
 d = \begin{bmatrix} d(1) \\ d(2) \end{bmatrix}
 \end{cases}$$

$$E[\theta | d(1), d(2)] = [\sigma_{\theta_1} \quad \sigma_{\theta_2}] \frac{1}{\sigma^2} \begin{bmatrix} \sigma_{22} & -\sigma_{12} \\ -\sigma_{12} & \sigma_{11} \end{bmatrix} \begin{bmatrix} d(1) \\ d(2) \end{bmatrix} =$$

$$= \frac{1}{\sigma^2} \left[\sigma_{\theta_1} \frac{\sigma_{22}}{\sigma_{11}} - \sigma_{\theta_2} \frac{\sigma_{21}}{\sigma_{11}} \quad \sigma_{\theta_2} - \sigma_{\theta_1} \frac{\sigma_{12}}{\sigma_{11}} \right] \begin{bmatrix} d(1) \\ d(2) \end{bmatrix} =$$

$$= \frac{1}{\sigma^2} \left(\sigma_{\theta_1} \frac{\sigma_{22}}{\sigma_{11}} - \sigma_{\theta_2} \frac{\sigma_{21}}{\sigma_{11}} \right) d(1) + \frac{1}{\sigma^2} \left(\sigma_{\theta_2} - \sigma_{\theta_1} \frac{\sigma_{12}}{\sigma_{11}} \right) d(2)$$

where

$$\sigma^2 = \sigma_{22} + \frac{\sigma_{21}^2}{\sigma_{11}}$$

Then, using innovation ...

$$\begin{aligned}
 E[\theta | d(1), d(2)] &= E[\theta | d(1)] + \underbrace{\frac{1}{\sigma^2}}_{= \frac{1}{\sigma_{ee}}} \underbrace{\left(\sigma_{y2} - \sigma_{y1} \frac{\sigma_{21}}{\sigma_{11}} \right)}_{= \sigma_{ee}} \underbrace{\left[d(2) - E[d(2) | d(1)] \right]}_{= e} = \\
 &= E[\theta | d(1)] + \frac{\sigma_{ee}}{\sigma_{ee}} e = \\
 &= E[\theta | d(1)] + E[\theta | e] = E[\theta | d(1)] + E[\theta | d(2)]
 \end{aligned}$$



$$E[\theta | d(1), d(2)] = E[\theta | d(1)] + E[\theta | d(2)]$$

Given a control law eq

$$\begin{cases} x(t+1) = Ax(t) + Bu(t) \\ y(t) = Cx(t) + Du(t) \end{cases}$$

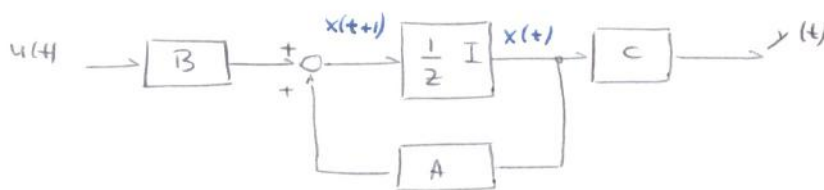
$D=0$ \forall system physically realizable

we have:

- A, B, C known
- y measured (disturbed without noise)
- $x(t+1)$ unknown

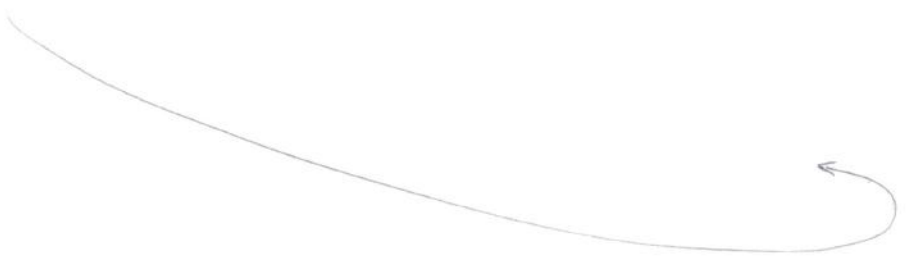
Goal is to estimate $\hat{x}(t) \approx x(t) \forall t$

Then, using Z transform and diagram blocks

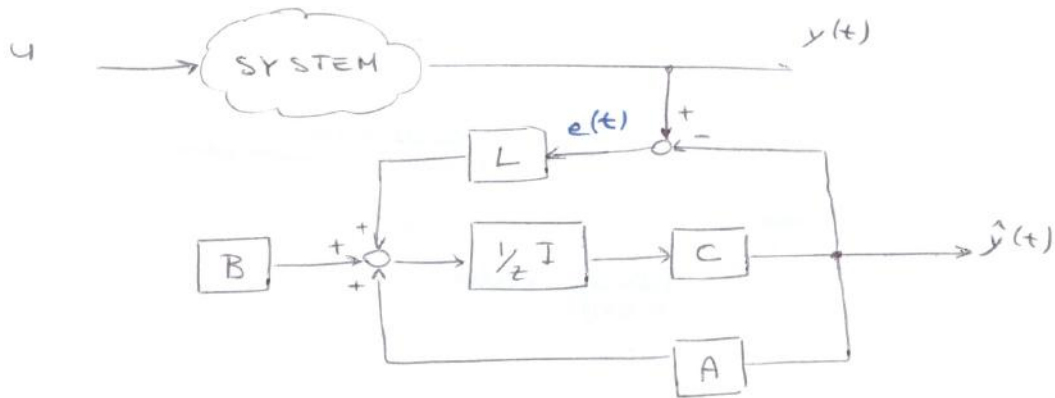


this is the system where $\hat{x}(t)$ is to be estimated

How to estimate $\hat{x}(t)$?



IDEA 2:



NOTE

$e(t) = y(t) - \hat{y}(t)$ has the same function as Bayesian estimate

So... According to block diagram

$$\tilde{x}(t+1) = (A - LC) \tilde{x}(t)$$

So, now, to guarantee $\tilde{x}(t) \rightarrow 0$ is needed only

$$|\lambda_i \{(A - LC)\}| < 1 \quad \forall i$$

⇒ Possible to set on L to fix the eigenvalues

⇓
All system possible can be estimated

Architecture: use L block to

bypass $|\lambda_i \{A\}| < 1$

If system fully observable

⇒

$$\rho(M_0) = m, \quad M_0 = [C \quad CA \quad CA^2 \dots]^T$$

L matrix can be designed to get

$$|\lambda_i(A - LC)| < 1 \quad \forall i$$

NOTE 2

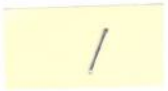
$$\bar{v}_1 = 0, \bar{v}_2 = 0 \Rightarrow \bar{x}(t) = 0, \bar{y}(t) = 0 \quad \forall t$$



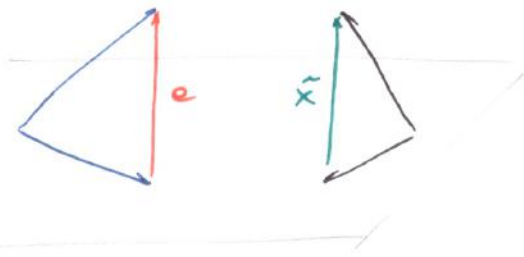
$$\hat{x}(N+r|N) = E[x(N+r)|y^N] = \sum_{x(N+r)y^N} \sum_{y^N}^{-1} y^N$$

Bayes Application

↑
inverting matrix
⇓
Hedy task



Considering e and \tilde{x} on the space:



$$e \perp \mathcal{H}, \tilde{x} \perp \mathcal{H} \Rightarrow e \parallel \tilde{x} \Rightarrow \underline{\underline{e, \tilde{x} \text{ linearly dependent}}}$$



Optimal estimate for $x(N+1)$ based on $y^N = \begin{bmatrix} y^N \\ y^{N-1} \\ \vdots \\ y^1 \end{bmatrix}$ is:

$$\hat{x}(N+1|N) = E[x(N+1) | y^N] =$$

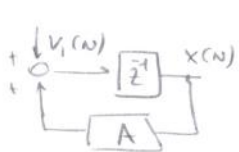
$$= E[x(N+1) | y^{N-1}, y^N] =$$

$$= \underset{\uparrow}{E[x(N+1) | y^{N-1}]} + E[x(N+1) | e(N)]$$

BAYES ESTIMATE

Then

$$E[x(N+1) | y^{N-1}] = E[Ax(N) + v_1(N) | y^{N-1}] =$$



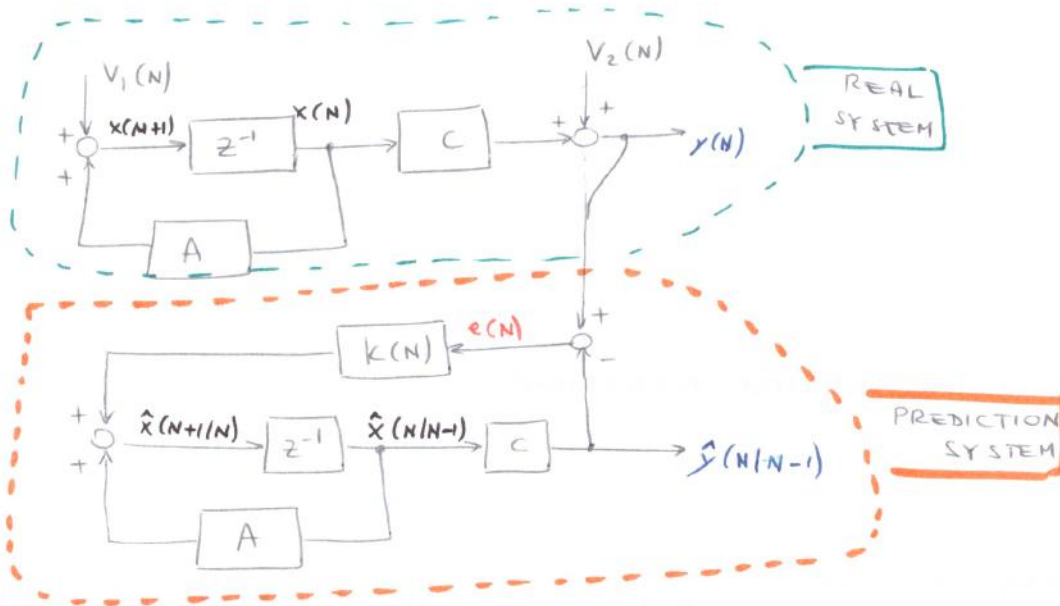
using Block Diagram

$$= AE[x(N) | y^{N-1}] + E[v_1(N) | y^{N-1}] =$$

$$= AE[x(N) | y^{N-1}] = A \hat{x}[N|N-1]$$

$$E[x(N+1) | e(N)] = \underbrace{\sum_{x(N+1), e(N)}^{-1}}_{K(N)} e(N) = K(N) e(N)$$

Then a block diagram could be:



REMARKS

- $k(N)$ involves prediction error variance $P(N)$ which changes during time
 \Downarrow
 $k(N)$ is time variant.
- KALMAN is the optimal linear prediction (based itself on BAYES)
- If
 - V_1, V_2 individually and jointly Gaussian $\forall t$
 - $x(t=0) \sim$ Gaussian
 - S linear

$\Rightarrow x(N+1), y^w$ jointly G. and individually G.



Bayesian Estimate of $x(N+1)$ is optimal

\Downarrow
 KALMAN predictor is the best

MULTI STEP KALMAN PREDICTOR

↓
used for $x(N+r)$, $r > 1$

Given the system as before the best optimal estimator for $x(N+r)$ given y^N is:

$$\begin{aligned} \hat{x}(N+r|N) &= E[x(N+r)|y^N] = \\ &= E[Ax(N+r-1) + v_1(N+r-1) | y^N] = \\ &= A E[x(N+r-1) | y^N] + E[v_1(N+r-1) | y^N] = \\ &= A \hat{x}(N+r-1|N) \end{aligned}$$

$\leftarrow = 0$ due white noise

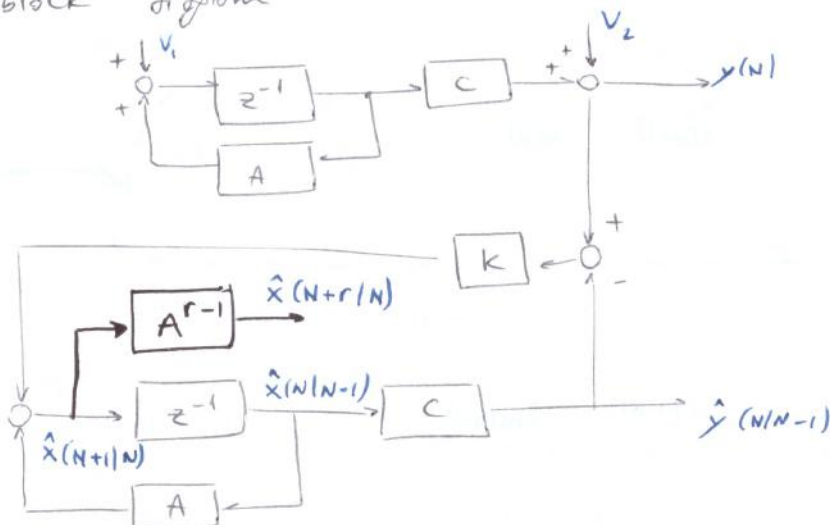
if

$$\begin{cases} \hat{x}(N+r+1|N) = A \hat{x}(N+r|N) \\ \hat{x}(N+r|N) = A \hat{x}(N+r-1|N) \dots \end{cases} \Rightarrow \hat{x}(N+r|N) = A^{r-1} \hat{x}(N+1|N)$$

$$\hat{x}(N+r|N) = A^{r-1} \hat{x}(N+1|N)$$

Solution is based on the next step KALMAN prediction

In block diagram



05/04/2017

KALMAN FILTER

Filtering data given measurements

IDEA

compute filtered data as a prediction summed something else

$$\text{Filtering} = \text{prediction} + (?)$$

Discovering (?)

Let's assume a LTI system Σ

$$\Sigma: \begin{cases} x(t+1) = Ax(t) + v_1(t) \\ y(t) = Cx(t) + v_2(t) \end{cases} \quad \text{where } \begin{aligned} v_1 &\sim WN(0, V_1) \\ v_2 &\sim WN(0, V_2) \\ x(t=1) &\sim (\bar{x}_1, P_1) \end{aligned}$$

Goal is estimate $x(N)$ as $\hat{x}(N|N)$

Optimal estimate is $\hat{x}(N|N) = E[x(N)|y^N] = E[x(N)|y^{N-1}, y(N)]$

Applying Bayes Recursive estimate

$$\hat{x}(N|N) = E[x(N)|y^{N-1}] + E[x(N)|e(N)] - E[x(N)]$$

$$E[x(N)|y^{N-1}] = \hat{x}(N|N-1)$$

$$E[x(N)|e(N)] = \sum_{x(N), e(N)} x(N) \sum_{e(N)}^{-1} e(N)$$

innovation $e(N) = y(N) - \hat{y}(N|N-1)$

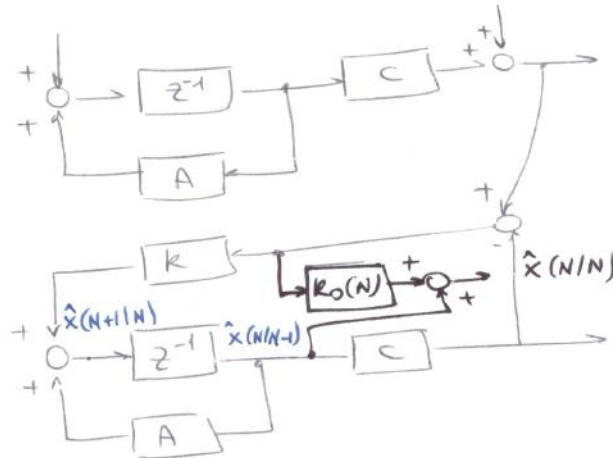
It's ok but should be better to have KALMAN parameters in order to have

$$\text{Filtering} = \text{Prediction} + (?)$$

$$\begin{aligned}
 V[\hat{x}(N)] &= V[x(N)] - V[\hat{x}(N/N-1)] = \\
 &= P(N) - P(N)C^T [C P(N)C^T + V_z]^{-1} C P(N) \\
 &\quad \Downarrow \\
 V[x(N) - \hat{x}(N/N)] &\leq P(N) = V[x(N)]
 \end{aligned}$$

Variance is getting smaller and smaller
 Filtering would be more accurate.

Filtering in block diagram



PREDICTOR / CORRECTOR ONE-STEP K.F.

provides reliable formulation to perform the $\hat{x}(N|N-1)$ to $\hat{x}(N+1|N)$ transitions.

During the prediction is also taking place a correction

performed in 2 steps

STEP ①: KALMAN FILTER

$$\hat{x}(N|N) = \hat{x}(N|N-1) + k_0(N)e(N)$$

STEP ②: COMPUTING $\hat{x}(N+1|N)$ using $\hat{x}(N|N)$

$$\hat{x}(N+1|N) = A\hat{x}(N|N) + Bu(N)$$

execution in time

$$k_0(N) = P(N)C^T [CP(N)C^T + V_2]^{-1}$$

$$P_0(N) = [I_m - k_0(N)C] P(N)$$

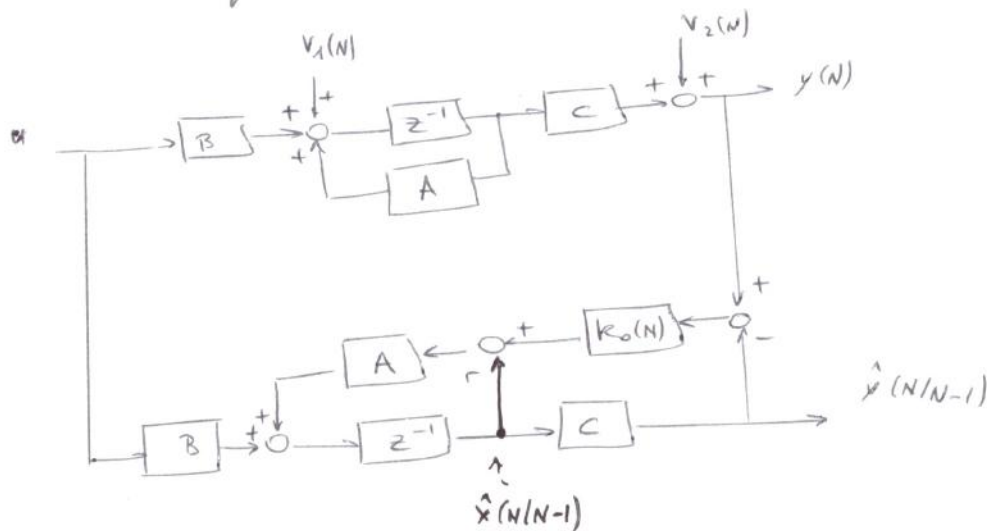
$$e(N) = y(N) - C\hat{x}(N|N-1)$$

$$\hat{x}(N|N) = \hat{x}(N|N-1) + k_0(N)e(N)$$

$$P(N+1) = AP(N)A^T + V_1$$

$$\hat{x}(N+1|N) = A\hat{x}(N|N) + Bu(N)$$

In block diagram



prediction error : $\tilde{x}(N+1) = x(N+1) - \hat{x}(N+1|N)$

$$x(N+1) = Ax(N) + v_1(N)$$

$$\hat{x}(N+1|N) = A \hat{x}(N|N-1) + \bar{K} e(N) =$$

$$= A \hat{x}(N|N-1) + \bar{K} [y(N) - \hat{y}(N|N-1)] =$$

$$= A \hat{x}(N|N-1) + \bar{K} [Cx(N) + v_2(N) - C \hat{x}(N|N-1)] =$$

$$= (A - \bar{K}C) \hat{x}(N|N-1) + \bar{K} C x(N) + \bar{K} v_2(N)$$

$$\tilde{x}(N+1) = Ax(N) + v_1(N) - (A - \bar{K}C) \hat{x}(N|N-1) - \bar{K} C x(N) - \bar{K} v_2(N) =$$

$$= (A - \bar{K}C) x(N) + v_1(N) - (A - \bar{K}C) \hat{x}(N|N-1) - \bar{K} v_2(N) =$$

$$= (A - \bar{K}C) [x(N) - \hat{x}(N|N-1)] + v_1(N) - \bar{K} v_2(N) =$$

$$\tilde{x}(N+1) = (A - \bar{K}C) \tilde{x}(N) + v_1(N) - \bar{K} v_2(N)$$

$$\tilde{x}(N+1) = (A - \bar{K}C) \tilde{x}(N) + v_1(N) - \bar{K} v_2(N)$$

↑ prediction error @ N+1 ↑ prediction error @ N

diff. eq. returning behavior of \tilde{x}

Considering a free-noise environment $\Rightarrow v_1 = v_2 = 0 \quad \forall i$

$$\tilde{x}(N+1) = (A - \bar{K}C) \tilde{x}(N)$$

computing a recursive formula $\tilde{x}(N) = (A - \bar{K}C)^{N-1} \tilde{x}(1)$

we want an error $\tilde{x}(n)$ getting smaller and smaller

$$|\lambda_i(A - \bar{K}C)| < 1 \quad \forall \lambda_i$$

looks like the internal stability

NON LINEAR KALMAN FILTERING

05/11/2017

Given S a non linear system

$$S: \begin{cases} x(t+1) = f(t, x(t), u(t)) \\ y(t) = h(t, x(t), u(t)) \end{cases} \quad \text{where } f, h \text{ non linear functions, known.}$$

It's set a nominal movement $\bar{x}(\cdot)$, considered as a standard behavior obtained setting a nominal input $\bar{u}(\cdot)$

If $\bar{x}(\cdot)$ is considered as a reference for other movements then can be defined:

Comparison between theoretical movements and real ones

- State perturbation $\delta x(t) = x(t) - \bar{x}(t)$
- Input perturbation $\delta u(t) = u(t) - \bar{u}(t)$
- Output perturbation $\delta y(t) = y(t) - \bar{y}(t)$

Exploiting perturbations is possible to linearize the system S as:

$$S_L: \begin{cases} \delta x(t+1) = A(t) \delta x(t) + B(t) \delta u(t) \\ \delta y(t) = C(t) \delta x(t) + D(t) \delta u(t) \end{cases}$$

$$A(t) = \left. \frac{\partial f}{\partial x} \right|_{\substack{x(t) = \bar{x}(t) \\ u(t) = \bar{u}(t)}}$$

$$B(t) = \left. \frac{\partial f}{\partial u} \right|_{\substack{x(t) = \bar{x}(t) \\ u(t) = \bar{u}(t)}}$$

$$C(t) = \left. \frac{\partial h}{\partial x} \right|_{\substack{x(t) = \bar{x}(t) \\ u(t) = \bar{u}(t)}}$$

$$D(t) = \left. \frac{\partial h}{\partial u} \right|_{\substack{x(t) = \bar{x}(t) \\ u(t) = \bar{u}(t)}}$$

of course δ must be relatively small to provide a good linear approximation

Jacobian Matrices

If A, B, C, D constant \Rightarrow system is in equilibrium

since the S_L is linear is possible to apply kalman predictor

$$\bar{K}: \begin{cases} \hat{x}(N+1|N) = \bar{A}(N)\hat{x}(N|N-1) + \bar{k}(N)e(N) \\ \hat{y}(N|N-1) = \bar{C}(N)\hat{x}(N|N-1) \\ e(N) = y(N) - \hat{y}(N|N-1) \end{cases}$$

where

$$\begin{cases} \bar{K}(N) = \bar{A}(N)P(N)\bar{C}(N)^T [\bar{C}(N)P(N)\bar{C}(N)^T + V_2]^{-1} \\ P(N+1) = \bar{A}(N)P(N)\bar{A}(N)^T + V_1 - \bar{K}(N) [\bar{C}(N)P(N)\bar{C}(N)^T + V_2] \bar{K}(N)^T \end{cases}$$

Predicted $\hat{x}(N+1|N) \neq x(N+1)$

↳ definition

$$\hat{x}(t) = x(t) - \bar{x}(t) \Rightarrow x(t) = \bar{x}(t) + \hat{x}(t)$$

⇓

$$\hat{x}(N+1|N) = \bar{x}(N+1) + \hat{x}(N+1|N)$$

↑
Given by kalman predictor

⚠ P.A.

Kalman is predicting the state perturbation and not the state.

Estimated state can be obtained by linearized kalman prediction \hat{x}

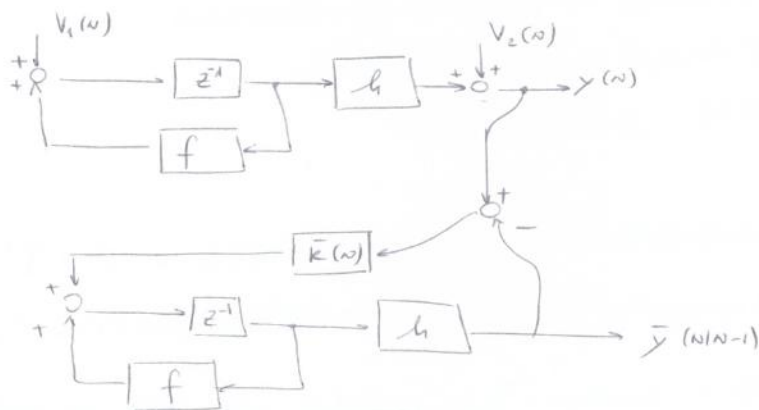
$$\hat{x}_k : \begin{cases} \hat{x}(N+1|N) = f(N, \hat{x}(N|N-1)) + \bar{K}(N)e(N) \\ \hat{y}(N|N-1) = h(N, \hat{x}(N|N-1)) \\ e(N) = y(N) - \hat{y}(N|N-1) \end{cases}$$

$\bar{K}(N)$ depends by $\bar{A}(N), \bar{C}(N)$ for the linearized system

computed previously around nominal movement

Real state is not taken in account

High estimation error \Rightarrow This strategy is not used at all



Extended KALMAN predictor is used in Robust Control System

In robust control

$$\begin{cases} x_1(t+1) = x_2(t) x_1(t) + v_{11}(t) \\ \dots \\ x_2(t+1) = x_2(t) + v_{12}(t) \end{cases}$$

x_2 should be a piecewise constant

Non-linear

using extended kalman estimate

KALMAN THEORY LIMIT :

- Good solution strongly depends by assumptions (a priori information)



Difficult to find out which assumptions are correct or not

